

To illustrate how a $Z-\theta$ transmission line chart can be used to determine the transmission characteristics of a sandwich, an example will be considered. For this example, the polarization is perpendicular to the plane of incidence, $\lambda_0 = 3.2 \text{ cm} = 1.26 \text{ inch}$, $\theta_0 = 45^\circ$, $d_1 = 0.05 \text{ inch}$, $\mu_1 = 1$, $\epsilon_1 = 6$, $d_2 = 0.21 \text{ inch}$, $\mu_2 = 0.8(1+j0.05)$, $\epsilon_2 = 4(1-j0.2)$, $d_3 = 0.07 \text{ inch}$, $\mu_3 = 1$, and $\epsilon_3 = 4$. Now $\eta_{0\perp} = \eta_{1\perp} = 1.414$, $\eta_{1\perp} = 0.426$, $\eta_{2\perp} = 0.481/7.9^\circ$, $\eta_{3\perp} = 0.535$, $\alpha_1 = \alpha_3 = 0$, $\alpha_2 = 0.721$ nepers per inch, $\beta_1 = (2\pi)(1.861)$ radians per inch, $\beta_2 = (2\pi)(1.317)$ radians per inch, and $\beta_3 = (2\pi)(1.485)$ radians per inch.

The point $Z_1 = \eta_{0\perp}/\eta_{1\perp} = 3.317$ is plotted on a $Z-\theta$ chart, as in Fig. 2 (below). The

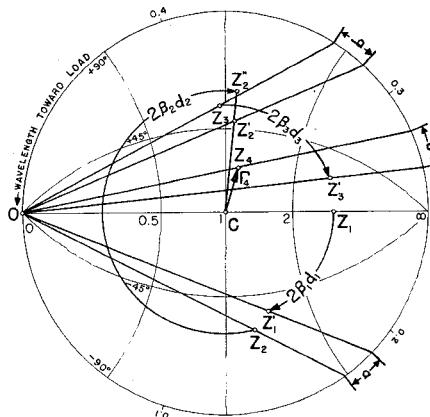


Fig. 2—Graphical construction for example.

point Z_1 is rotated through the angle $2\beta_1 d_1$ or $(1.861)(0.05)\lambda = 0.093\lambda$ to obtain $Z_1' = 1.40/-54^\circ$. The point $Z_2 = (\eta_{1\perp}/\eta_{2\perp})Z_1' = 1.24/-61.9^\circ$ is plotted. The point Z_2 is rotated through the angle $2\beta_2 d_2$ or $(1.317)(0.21)\lambda = 0.278\lambda$ to obtain Z_2'' . The distance $CZ_2'' e^{-2\alpha_2 d_2} = CZ_2'' e^{-0.0303} = 0.739 CZ_2''$ is measured from C along the line CZ'' to locate the point $Z_2' = 1.07/49^\circ$. The point $Z_3 = (\eta_{2\perp}/\eta_{3\perp})Z_2' = 0.96/56.9^\circ$ is plotted. The point Z_3 is rotated through the angle $2\beta_3 d_3$ or $(1.485)(0.07)\lambda = 0.104\lambda$ to obtain $Z_3' = 2.98/25^\circ$. The point $Z_4 = (\eta_{3\perp}/\eta_{0\perp})Z_3' = 1.13/25^\circ$ is plotted.

The magnitude of the voltage reflection coefficient is $\overline{CZ_4}/(\text{radius of chart}) = 0.23$ Now

$$|E_4^+| = \frac{\overline{OC}}{\overline{OZ_1}} \frac{\overline{OZ_1}}{\overline{OZ_2}} \frac{\overline{OZ_2}}{\overline{OZ_3}} \frac{\overline{OZ_3}}{\overline{OZ_4}} e^{\alpha_1 d_1 + \alpha_2 d_2 + \alpha_3 d_3} = 1.10$$

and the ratio of power transmitted to power incident is 0.826. The angle of E_4^+ is

$$\begin{aligned} a - b - c + \beta_1 d_1 + \beta_2 d_2 + \beta_3 d_3 \\ = 360^\circ(0.014 - 0.016 - 0.016 + 0.093 \\ + 0.278 + 0.104) \\ = 164.5^\circ, \end{aligned}$$

and the increase in phase retardation due to the sandwich is

$$\Phi = 164.5^\circ - (360^\circ/1.26)(0.33) \cos 45^\circ = 9.78^\circ.$$

A graphical construction for determining the transmission characteristics of a sandwich has the advantage that the effect of the individual layers is presented in visual form. Ways to vary the parameters to obtain a desired result may be suggested by a study of the chart.

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Measurement of Reflection Coefficients through a Lossless Network

Often it is not convenient to measure a reflection coefficient directly. In some such

cases, the following procedure may be used. The arrangement of components shown in Fig. 1 will be considered. It will be assumed that lines Nos. 1 and 2 are lossless. These lines may be transmission lines, waveguides, or one of each.

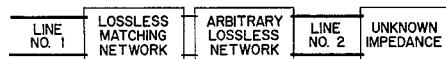


Fig. 1—Arrangement for measuring reflection coefficients.

First, a matched load is connected to line No. 2 and the matching network is adjusted until the reflection coefficient in line No. 1 is zero. Next, the matched load is replaced by a short circuit. The magnitude of the reflection coefficient in line No. 1 should be one. This should be checked experimentally. A voltage null in line No. 1 is located and designated the *short-circuit point*. Finally, the short circuit is replaced by the unknown impedance. The reflection coefficient Γ_1 in line No. 1 is measured relative to the short-circuit point. The reflection coefficient Γ_2 in line No. 2 related to the point where the short circuit was connected is the same as Γ_1 .

The theoretical basis for this procedure is the well-known equation

$$\Gamma_1 = \frac{a\Gamma_2 + b}{c\Gamma_2 + d}.$$

Since $\Gamma_1 = 0$ when $\Gamma_2 = 0$, $b = 0$. Since the various components are lossless, $\Gamma_1 = 1$ when $\Gamma_2 = 1$. Consequently, $|c\Gamma_2 + d|$ is equal to a constant for all values of Γ_2 such that $\Gamma_2 = 1$. This requires that $c = 0$. Now $\Gamma_1 = (a/d)\Gamma_2$, where $a/d = 1$. When Γ_1 is referred to the short-circuit point, $a/d = 1$ and $\Gamma_1 = \Gamma_2$.

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